

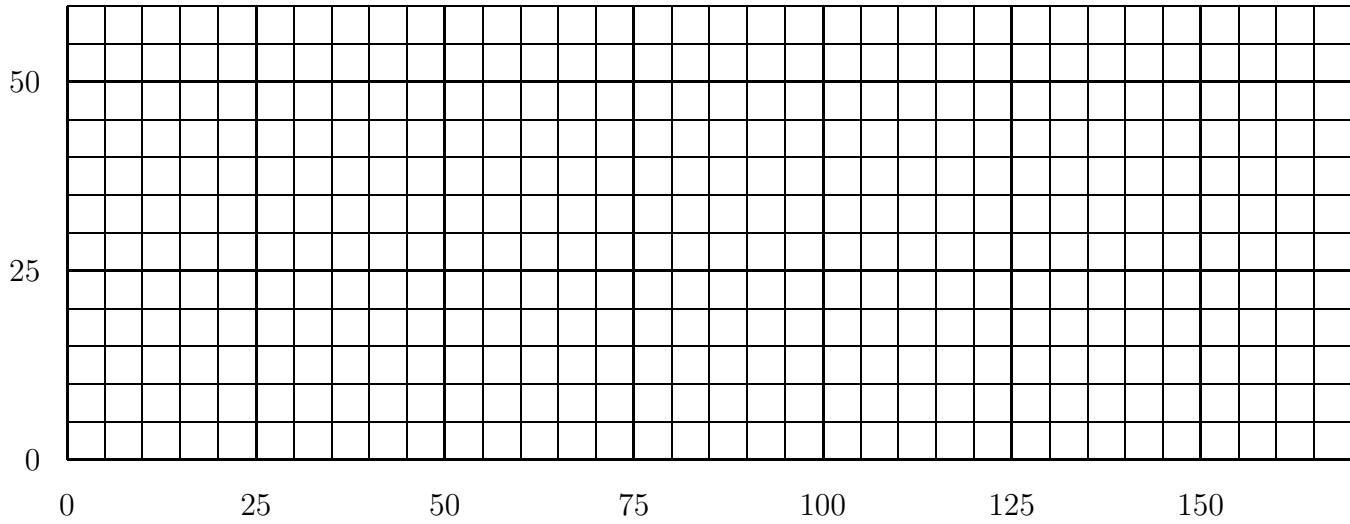
EE.351: Spectrum Analysis and Discrete-Time Systems  
FINAL EXAM, 9:00AM–12:00PM, December 12, 2003 (**closed book**)  
Examiner: Ha H. Nguyen

*Note: There are six questions. All questions are of equal value but not necessarily of equal difficulty. Full marks shall only be given to solutions that are properly explained and justified.*

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1. (*FS of a continuous-time signal*) Consider a half-wave rectifier whose output  $y(t)$  is related to the input  $x(t)$  as follows:  $y(t) = \begin{cases} x(t), & \text{if } x(t) \geq 0 \\ 0, & \text{if } x(t) < 0 \end{cases}$ . Let  $x(t) = \cos(2\pi t)$ .

- [3] (a) Neatly sketch the input  $x(t)$  and the output  $y(t)$ . What are the fundamental periods of  $x(t)$  and  $y(t)$ ?



- [5] (b) Determine the Fourier series coefficients of the output  $y(t)$ .

- [2] (c) What are the amplitudes of the dc components of the input and output signals, respectively?

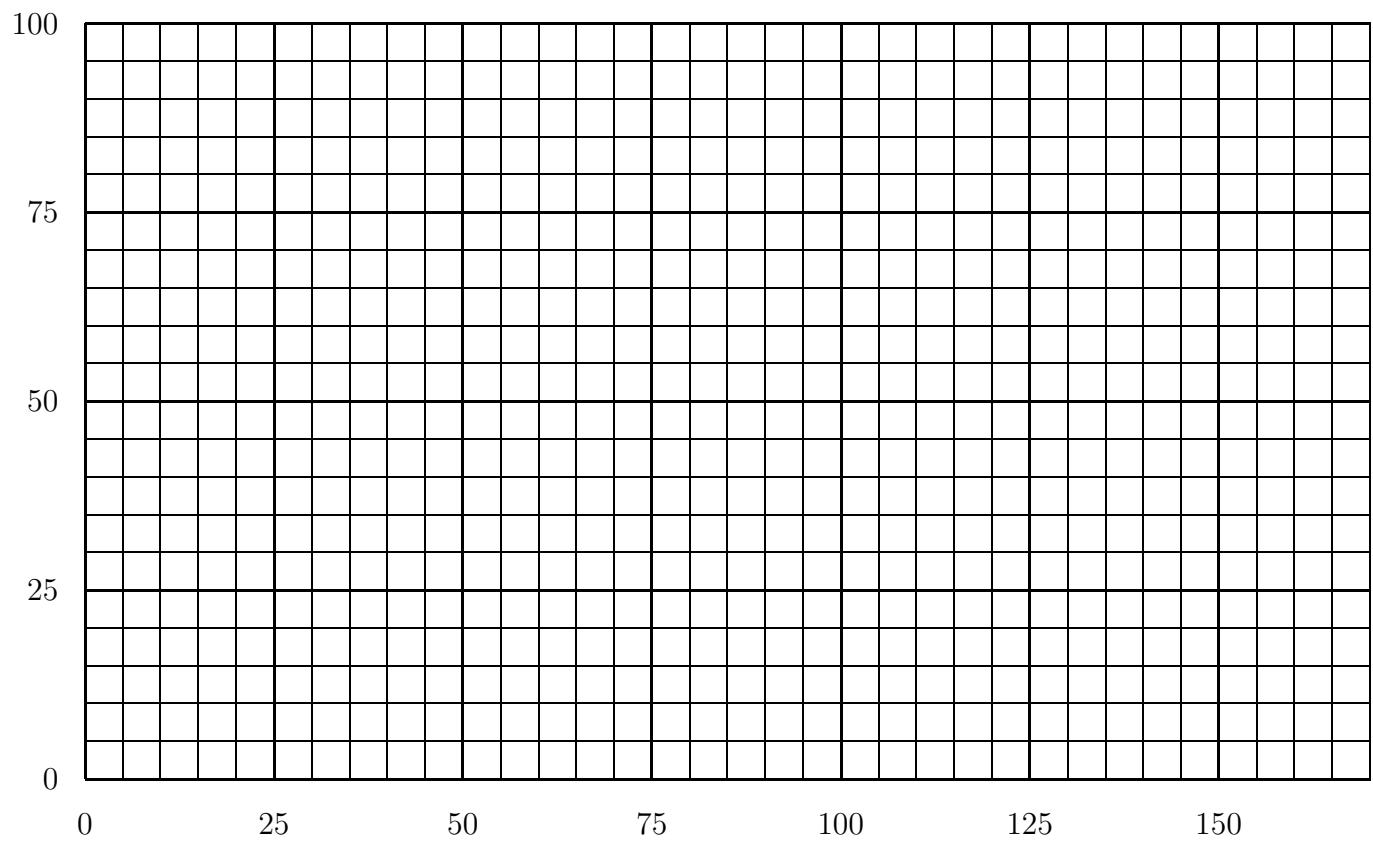
2. (*FS of a discrete-time signal*) Consider the following discrete-time periodic signal

$$x[n] = 2 + 2 \cos\left(\frac{\pi}{3}n - \frac{\pi}{4}\right) + \sin\left(\frac{2\pi}{3}n - \frac{\pi}{6}\right)$$

[2] (a) What are the fundamental frequency and fundamental period of  $x[n]$ ?

[5] (b) Find the Fourier series coefficients for  $x[n]$ .

- [3] (c) Plot the magnitude and phase spectrum of  $x[n]$  over two periods.

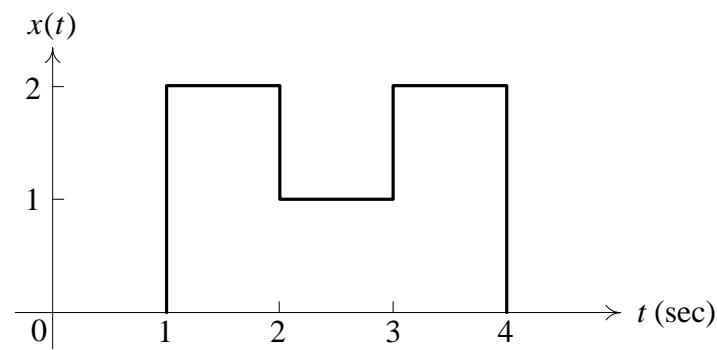


3. (Fourier Transform) The signal  $x(t)$  has Fourier transform  $X(j\omega)$ .

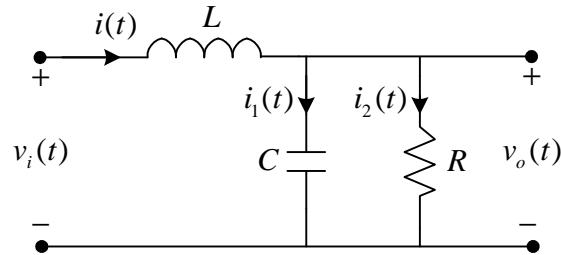
- [2] (a) Show that the Fourier transform of  $dx(t)/dt$  is  $(j\omega)X(j\omega)$ .

- [2] (b) Show that the Fourier transform of  $x(t - t_0)$  is  $e^{-j\omega t_0} X(j\omega)$ , where  $t_0$  is a constant.

- [6] (c) Find  $X(j\omega)$  if  $x(t)$  is given below.



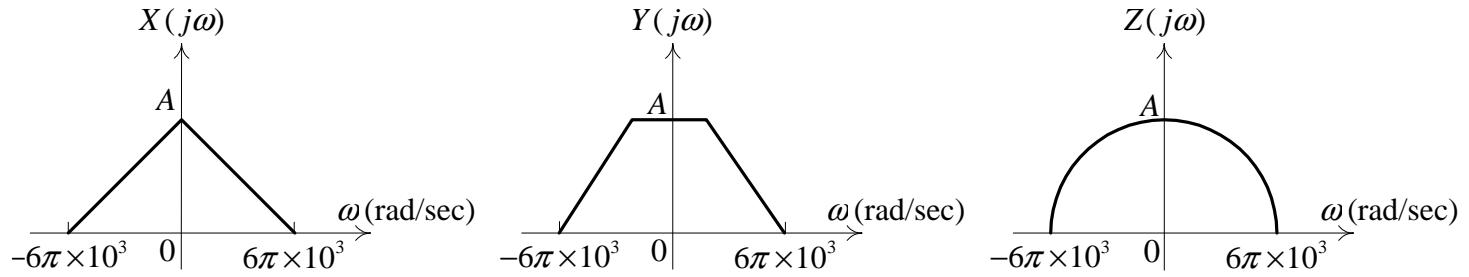
4. (Frequency response of an RLC filter) Consider the following circuit:



- [5] (a) Show that the frequency response of the circuit is given by  $H(j\omega) = \frac{1}{1 - \omega^2 LC + j\omega L/R}$

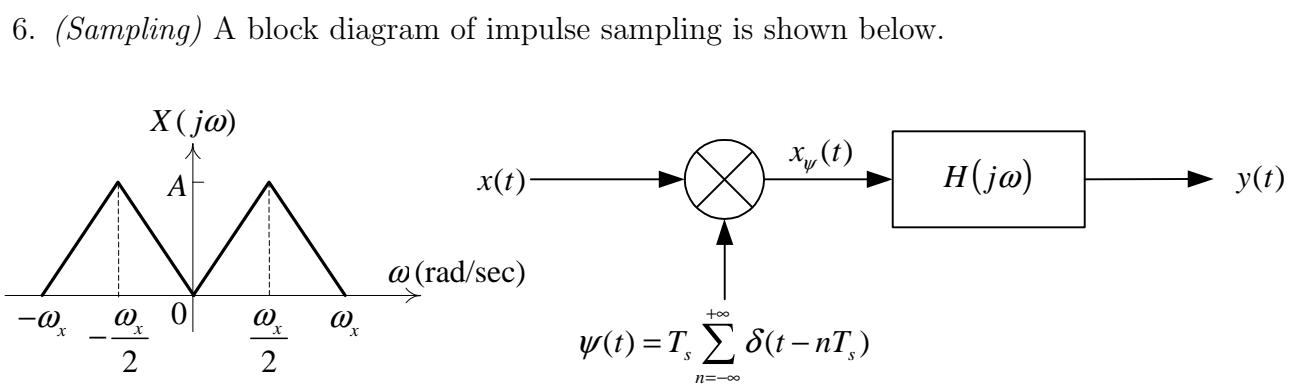
- [5] (b) Let  $\omega_c = \sqrt{\frac{1}{LC}}$  and  $L = 2R^2C$ . Show that the magnitude frequency response of the circuit can be written as  $|H(j\omega)| = \frac{1}{\sqrt{1 + (\omega/\omega_c)^4}}$ . Is the circuit a low-pass or high-pass filter? Explain your answer.

5. (*Amplitude modulation*) The manager of your division asked you to design a communication system to transmit three signals  $x(t)$ ,  $y(t)$  and  $z(t)$  simultaneously over a radio channel. The spectra of three signals are shown below. You bought a radio frequency spectrum from  $650$  kHz to  $668$  kHz (i.e., from  $2\pi \times 650 \times 10^3$  rad/sec to  $2\pi \times 668 \times 10^3$  rad/sec) and decided to use amplitude modulation.



- [5] (a) Clearly present a block diagram of the transmitter to your manager. Also show him the spectrum of the transmitted signal. Clearly identify all the relevant frequencies.

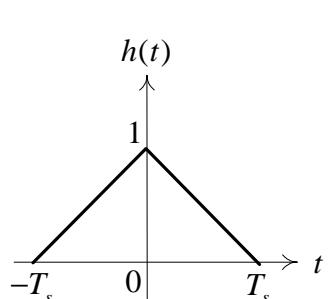
- [5] (b) To convince your manager that your design is valid, show and explain to him a block diagram of the receiver that can perfectly recover the signal  $y(t)$  (assume that there is no distortion introduced by the channel).



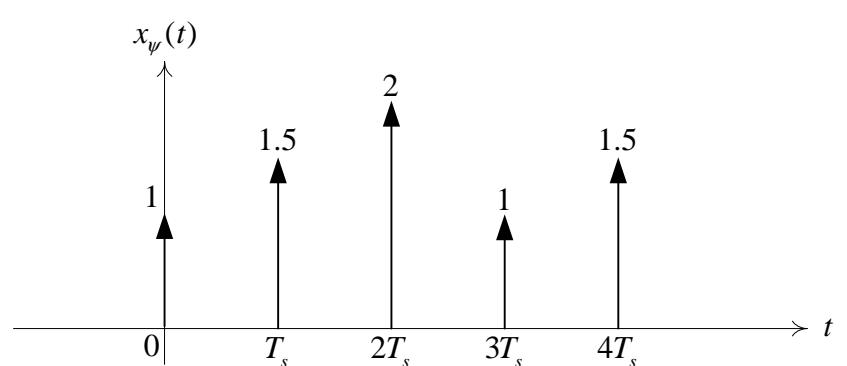
- [2] (a) If the sampling frequency is  $\omega_s = \frac{2\pi}{T_s} = \frac{3\omega_x}{2}$ , does aliasing occur? What is the minimum sampling frequency to prevent aliasing?

- [5] (b) Draw the spectrum of  $x_\psi(t)$  when  $\omega_s = \frac{3\omega_x}{2}$ . Also draw the spectrum of  $y(t)$  if  $H(j\omega)$  is an ideal low-pass filter with a cutoff frequency of  $\omega_c = \omega_x$ .

- [3] (c) Instead of the ideal low-pass filter, consider the filter in Figure (a) as a reconstruction filter. Find and plot the reconstructed signal  $y(t)$  if  $x_\psi(t)$  is as shown in Figure (b) over the interval  $[0, 4T_s]$ .



(a)



(b)

### Potentially Useful Facts:

- FS representation of DT periodic signals (DTFS):

$$x[n] = \sum_{k=<N>} a_k e^{jk\omega_0 n}, \quad a_k = \frac{1}{N} \sum_{n=<N>} x[n] e^{-jk\omega_0 n}$$

- FS representation of CT periodic signals (CTFS):

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}, \quad a_k = \frac{1}{T_0} \int_{T_0} x(t) e^{-jk\omega_0 t} dt$$

- FT of CT signals:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega, \quad X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt$$

- Fourier transform pairs/properties:

$$\begin{aligned} \cos(\omega_0 t) &\xleftrightarrow{\mathcal{FT}} \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0) \\ \delta(t) &\xleftrightarrow{\mathcal{FT}} 1 \\ x(t - t_0) &\xleftrightarrow{\mathcal{FT}} e^{-j\omega t_0} X(j\omega) \\ \frac{d}{dt} x(t) &\xleftrightarrow{\mathcal{FT}} j\omega X(j\omega) \\ \int_{-\infty}^t x(\tau) d\tau &\xleftrightarrow{\mathcal{FT}} \frac{1}{j\omega} X(j\omega) \quad (\text{Ignoring DC signal}) \\ \cos(\omega_0 t)x(t) &\xleftrightarrow{\mathcal{FT}} \frac{1}{2}X(j(\omega - \omega_0)) + \frac{1}{2}X(j(\omega + \omega_0)) \\ x(t)\psi(t) &\xleftrightarrow{\mathcal{FT}} \sum_{k=-\infty}^{\infty} X(j(\omega - k\omega_s)) \quad (\text{Impulse sampling}) \end{aligned}$$

- Identities:

$$\begin{aligned} e^{j\omega n} &= \cos(\omega n) + j \sin(\omega n) \\ e^{-j\omega n} &= \cos(\omega n) - j \sin(\omega n) \\ \cos(x) \cos(y) &= \frac{1}{2}[\cos(x+y) + \cos(x-y)] \end{aligned}$$

For my own survey, please indicate () which equations provided above are useful for you, i.e., the ones that you used and you might not remember if not provided. Thanks.